# Tripoli university Faculty of engineering EE department EE313 tutorial (uniform plane waves)

Notes:-

\*Perfect dielectric  $\Rightarrow$  **d**=0,  $\beta = \omega \sqrt{\mu} \in \mathcal{M} = \sqrt{\frac{M}{\epsilon}} \cdot (\sigma = 0)$ \*Lossy dielectric  $\Rightarrow \sigma \neq 0$ .

\* Nonmagnetic material  $\Rightarrow M_r = 1$ ,  $M = M_o$ .

\* Loss tangent =  $\frac{\sigma}{\omega \epsilon}$  or  $\frac{\epsilon''}{\epsilon i}$ .

\* To obtain the field in real time form :-

 $E(z,t) = Re \{\hat{E}(z)e^{j\omega t}\}$ 

#### Problem#1.

An electromagnetic wave in free space has a wavelength of 0.2 m. when this same wave enters a perfect dielectric, the wavelength changes to 0.09 m. Assuming that Mr=1, determine Er.

#### solution

In free space:  $\beta = \omega \sqrt{M_0 \epsilon_0}$   $\longrightarrow$  (1)

In the perfect dielectric:  $\beta = W \sqrt{ME} = W \sqrt{ME_r E_o} \longrightarrow (2)$ 

wavelength in free space  $\lambda = \frac{2\pi}{\beta} \longrightarrow (3)$  wavelength in the dielectric  $\lambda = \frac{2\pi}{\beta} \longrightarrow (4)$ 

Dividing equation (4) by equation (3):

$$\frac{0.09}{0.2} = \frac{\frac{2\pi}{\beta}}{\frac{2\pi}{\beta}} = \frac{\beta_{o}}{\beta} = \frac{\omega\sqrt{M_{o}\epsilon_{o}}}{\omega\sqrt{M_{o}\epsilon_{r}\epsilon_{o}}} = \frac{1}{\sqrt{\epsilon_{r}}}$$

#### Problem#2

In some region  $\sigma = 0.01$ ,  $E = 2E_0$ ,  $\mu = \mu_0$ . The magnitude of electric field at z = 0 is 100 V/m. At frequency of 100 MHz, write the instantaneous expression for E and H. At what depth the E field magnitude reduce to 1%.

Solution
$$\frac{\sigma}{\omega \epsilon} = 0.899$$

$$\alpha = \frac{\omega \sqrt{\mu \epsilon}}{\sqrt{2}} \left[ \sqrt{1 + \left(\frac{\sigma}{\omega \epsilon}\right)^2} - 1 \right]^2 = 1.23 \text{ Np/m}.$$

$$\beta = \frac{\omega \sqrt{\mu \epsilon}}{\sqrt{2}} \left[ \sqrt{1 + \left(\frac{\sigma}{\omega \epsilon}\right)^2} + 1 \right]^{\frac{1}{2}} = 3.209 \text{ rad/m}.$$

$$|\hat{m}| = \frac{\sqrt{\frac{M}{\epsilon}}}{1 + \left(\frac{\sigma}{\omega \epsilon}\right)^2}|_{\frac{1}{4}} = 229.9 \text{ S2}.$$

$$\widehat{\underline{M}} = \frac{1}{2} \tan^{-1} (\overline{\underline{\omega}} \underline{\epsilon}) = 21^{\circ}$$

$$E(\mathbf{Z}, t) = E_{m}^{+} e^{-\alpha Z} \cos(\mathbf{\omega} t - \beta Z)$$

$$= 100 e^{1.23Z} \cos(2\pi \times 100 \times 10^{6} t - 3.209Z)$$

$$H^{+}(\mathbf{Z}, t) = \frac{E_{m}^{+}}{|M|} e^{-\alpha Z} \cos(\omega t - \beta Z - \underline{\underline{M}})$$

$$= 0.435 e^{-1.237} \cos \left(\frac{3.2092\pi \times 100 \times 10^{6}t}{3.2097} - 21^{\circ}\right)$$

The magnitude of the E-field is  $E_m^+ e^{\alpha z}$ . this will reduce to 1% of 100 (or 1) at distance di-

$$100e^{-1.23d}$$
  
= 1  $\Rightarrow$  d= 3.744 m.

Problem#3

The electric field of a wave propagating in a nonmagnetic lossy dielectric is  $\hat{E}^{\dagger}(z) = \vec{a_x} \cdot 10\vec{e}^{72}$ ,  $\gamma = 3.93 + j4.018$  with a frequency of 20MHz. Find the magnetic field of the wave.

#### Solution:

To find H, we have to find  $\tilde{\eta}$ , but we don't have  $\epsilon$  and the loss tangent  $\tilde{\epsilon}$ .

$$\gamma = \alpha + j\beta = 3.93 + j4.018 \Rightarrow \alpha = 3.93, \beta = 4.018$$

$$\frac{\beta}{\alpha} = \frac{4.018}{3.93} = \frac{\omega\sqrt{M\epsilon} \left[\sqrt{1+\left(\frac{\epsilon''}{\epsilon'}\right)^2+1}\right]^{\frac{1}{2}}}{\frac{\omega\sqrt{M\epsilon}}{\sqrt{2}} \left[\sqrt{1+\left(\frac{\epsilon''}{\epsilon'}\right)^2-1}\right]^{\frac{1}{2}}}$$

$$\int 1 + \left(\frac{\varepsilon^{"}}{\varepsilon}\right)^{2} + 1 = 1.0453$$

$$\int 1 + \left(\frac{\varepsilon^{"}}{\varepsilon}\right)^{2} - 1$$

After some algebric manipulations:

$$\frac{\epsilon^{"}}{\epsilon^{"}} = 44.9$$

By substituting this value into the expression of xi-

$$3.93 = \frac{2\pi \times 20 \times 10^{6} \times \sqrt{4\pi \times 10^{7}} \times \epsilon_{r} \times 8.854 \times 10^{12}}{\sqrt{2}} \left[ \sqrt{1 + (44.9)^{2}} - 1 \right]^{\frac{1}{2}}$$

$$3.93 = 1.963 \sqrt{\epsilon_r} \implies \epsilon_r = 4$$

And now we can find m.

$$|\hat{M}| = \frac{\sqrt{M/\epsilon}}{\left[1 + \left(\frac{\epsilon}{\epsilon}\right)^2\right]^{\frac{1}{4}}} = 28.152$$

$$\angle \hat{n} = \pm \tan^{-1}(\frac{\epsilon}{\epsilon}) = 44.36^{\circ}$$

$$\hat{H}_{y}^{+}(z) = \frac{\hat{E}_{x}^{+}(z)}{\hat{m}}$$

$$= 0.3559 e^{-3.937} e^{-j(4.0187 + 44.36°)}$$

Note that the angle of M is nearly 45° because of the large loss tangent.

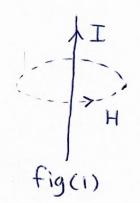


# Tripoli university Faculty of engineering EE department. EE313 Tutorial (Magnetic polarization)

### Problem (3-26)

Note the similarity between this problem and example (1-17).

Using right hand rule the H field of a stright current carrying Conductor is found as in fig(i).

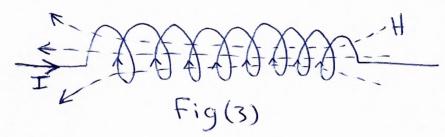


And for single loop by applying the same right hand rule, H will be as shown in fig(2).

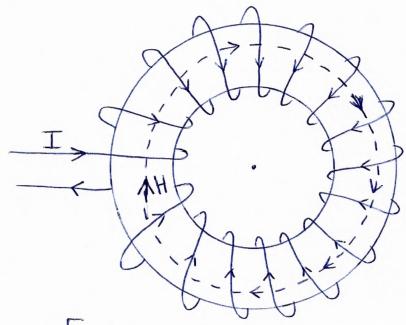


fig (2)

And for n turns solenoidal coil the magnetic field will be as in fig(3)

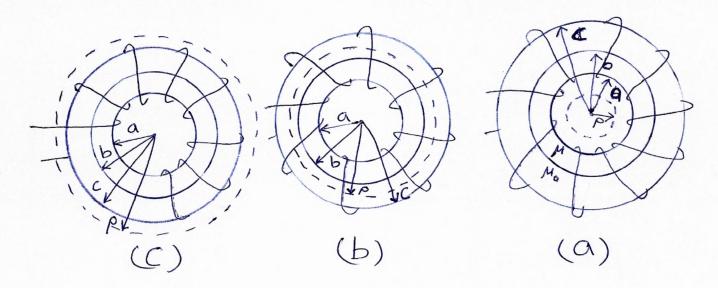


And if this solenoid is bent into a toroide we find that the H field inside the toroid must be in ap direction as shown in figure.



Fig(4), (note that Z-axis is directed out of the paper).

Now, for the toroid of our problem if we take ampere's path for P < a as shown in fig(50) there will be no current crossing the area bounded by the path and hence H = 0. For a < P < C (inside the toroid and no difference between the air and the magnetic material) amper's path is shown in Fig(5b)



Fig(5)

Here, the current I crosses the area bounded by the path n times. using Ampere's law:-

$$\int_{l} H_{\phi} \vec{a}_{\phi} \cdot dl = nI$$

$$\int_{0}^{2\pi} H_{\phi} \vec{a}_{\phi} \cdot Pd\phi \vec{a}_{\phi} = nI$$

$$H_{\phi} = \frac{nI}{2\pi \rho}, \quad \alpha < P < C$$

For P > C, the current I enter the area bounded by the path in opposite directions n times, hence the are cancelling each other in Ampere's law makes H = 0. (see fig 5c). H is the same in the magnetic material and in the air but B=MH is not the same since M for the magnetic material is Mr.Mo and M for the air is Mo.

For 
$$a < P < b$$
:
$$B = \overrightarrow{a_{\phi}} \frac{M_r M_o n I}{2\pi P}$$
For  $b < P < C$ :
$$\overrightarrow{B} = \overrightarrow{a_{\phi}} \frac{M_o n I}{2\pi P}$$

c) 
$$\vec{M} = \frac{\vec{B}}{M_0} - \vec{H}$$
.

For aLP<b:

$$\overrightarrow{M} = \left[\frac{M_r n I}{2\pi \rho} - \frac{n I}{2\pi \rho}\right] \overrightarrow{a_{\phi}} = \overrightarrow{a_{\phi}} \frac{n I (M_{r-1})}{2\pi \rho}$$

For b<Pcc:

$$\overrightarrow{M} = \frac{nI}{2\pi\rho} - \frac{nI}{2\pi\rho} = 0.$$

d) For acpcbi-
$$\vec{J}_{m} = \vec{\nabla} \times \vec{M} = \begin{vmatrix} \vec{ap} & \vec{ap} & \vec{ap} \\ \vec{p} & \vec{ap} & \vec{ap} \\ \vec{p} & \vec{p} & \vec{p} \end{vmatrix} = 0$$

$$0 \frac{nI(M-1)}{2\pi} 0$$

And now, we want to find Jsm (surface current density on the surface of the magnetic material due to polarization).

 $\overrightarrow{J}_{SM} = -\overrightarrow{n} \times \overrightarrow{N}$ ,  $\overrightarrow{n}$  is the normal vector on the surface.

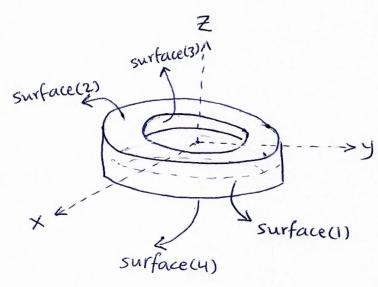


Fig (6)

For Surface (1) (n=ap):

$$\overrightarrow{J}_{sm} = -\overrightarrow{a_{p}} \times \overrightarrow{a_{\phi}} \frac{nI(\mu_{r-1})}{2\pi p} = -\overrightarrow{a_{z}} \frac{nI(\mu_{r-1})}{2\pi p}$$
(n is number of turns)

For surface(2) (n=az):

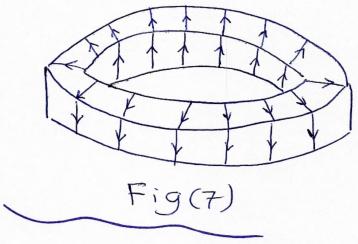
$$\vec{J}_{Sm} = -\vec{a}_{Z} \times \vec{a}_{\phi} \frac{nI(M_{r}-1)}{2\pi\rho} = \vec{a}_{\rho} \frac{nI(M_{r}-1)}{2\pi\rho}$$

$$\vec{J}_{sm} = \vec{a}_{p} \times \vec{a}_{\phi} \frac{nI(M_{r}-1)}{2\pi p} = \vec{a}_{z} \frac{nI(M_{r}-1)}{2\pi p}$$

For Surface (4) (
$$\vec{n} = -\vec{a}_z$$
)
$$\vec{J}_{sm} = +\vec{a}_z \times \vec{a}_\phi \frac{nI(M_r-1)}{2\pi\rho} = -\vec{a}_\rho \frac{nI(M_r-1)}{2\pi\rho}$$

Fig (7) shows a sketch of  $J_{sm}$  on the magnetic material:

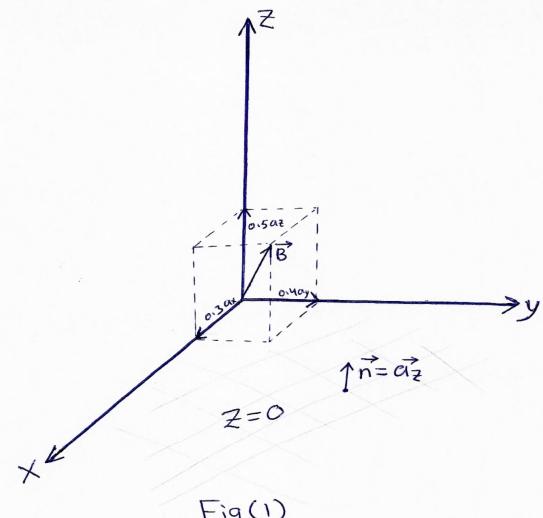
note that for surface (1) (P=b) and for surface (3) (P=a).



## Problem (3-27)

The plane Z=0 separates the space into two regions, Z>0 is air and Z<0 is a magnetic material with  $M_r=4$ .

 $\vec{B}_1 = 0.3 \, \vec{a}_X + 0.4 \, \vec{a}_Y + 0.5 \, \vec{a}_Z$  in air is sketched in Fig(1). note that the normal vector on this surface is  $\vec{a}_Z$ .



$$\overrightarrow{H}_1 = \frac{\overrightarrow{Bi}}{\cancel{M_0}} = \frac{0.3}{\cancel{M_0}} \overrightarrow{a_x} + \frac{0.4}{\cancel{M_0}} \overrightarrow{a_y} + \frac{0.5}{\cancel{M_0}} \overrightarrow{a_z}$$

From boundary conditions:

OFB

Bni=Bnz (normal components in each region are equal).

$$B_{12} = B_{22} = 0.5 \vec{a}_2$$

Ht1 = Ht2 (tangential components of H are equal).

$$\therefore \frac{0.3}{M_0} \vec{a_x} + \frac{0.4}{M_0} \vec{a_y} = H_{t2}$$

$$\vec{B}_2 = 1.2\vec{a}_x + 1.6\vec{a}_y + 0.5\vec{a}_z$$



The angle between  $\vec{B}_2$  and the normal on the surface  $\theta_2$  is equal to:

$$\cos \theta_{2} = \frac{\vec{a}_{2} \cdot \vec{B}_{2}}{|\vec{B}_{2}|} = \frac{0.5}{\sqrt{1.2^{2} + 1.6^{2} + 0.5^{2}}} \Rightarrow \theta_{2} = 76^{\circ}.$$
  
Similarly  $\theta_{1} = 45^{\circ}.$ 

# Problem (3-29)

$$\overrightarrow{E}_{1} = -15\overrightarrow{a_{x}} + 20\overrightarrow{a_{y}} + 30\overrightarrow{a_{z}}$$

$$\overrightarrow{D}_{1} = \xi_{0} \overrightarrow{E}_{1} = -15 \xi_{0} \overrightarrow{a}_{x} + 20 \xi_{0} \overrightarrow{a}_{y} + 30 \xi_{0} \overrightarrow{a}_{z}$$

Dni

From boundary conditions  $D_{n_1} = D_{n_2} = 30 \in \vec{a_z} \implies E_{n_2} = \frac{30 \in \vec{a_z}}{4 \in \vec{a_z}} = 7.5 \vec{a_z}$   $E_{t_1} = E_{t_2} = -15 \vec{a_x} + 20 \vec{a_y}$   $\vec{E_z} = -15 \vec{a_x} + 20 \vec{a_y} + 7.5 \vec{a_z}$ 

$$\cos \theta_2 = \frac{\vec{a_2} \cdot \vec{E_2}}{|\vec{E_2}|} = \frac{7.5}{\sqrt{15^2 + 20^2 + 7.5^2}} \implies \theta_2 = 73.3^\circ$$

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